A direct method approach for data-driven inference of high accuracy adaptive phase-isostable reduced order models

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Abstract

Phase-amplitude reduction techniques have shown great promise for identifying analytically tractable reduced order models in applications involving strongly perturbed and strongly coupled oscillatory dynamical systems. However, efficient methods for inference of these reduced order models from data are still needed. In this work, a data-driven strategy for inference of the necessary terms comprising an adaptive phase-isostable reduced order model is proposed. This strategy requires no more data than the well-established direct method used for obtaining standard phase reduced models, i.e., the application of a series of pulse inputs and the subsequent examination of the relaxation to the underlying limit cycle. Illustrative examples are provided for a collection of numerical models where the resulting adaptive phase-isostable reduced order equations are substantially more accurate than models obtained using standard phase reduction.

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1. Introduction

Phase reduction is a well-established strategy for reducing the complexity and dimensionality of oscillatory dynamical systems [1–4]. Using this strategy, a system with a stable limit cycle of the general form

$$\dot{x} = F(x, u),$$

(1)

where $F$ determines the system dynamics, $x \in \mathbb{R}^N$ denotes the system state, and $u \in \mathbb{R}^M$ corresponds to an input can be represented according to

$$\dot{\theta} = \omega + Z(\theta)\mu(t).$$

(2)

Above, $\theta \in [0, 2\pi)$ encodes for oscillation timing, $\omega$ is the unperturbed natural frequency that results when taking $u(t) = 0$, and $Z(\theta)$ is a row vector, often referred to as the infinitesimal phase response curve, that characterizes the phase shifts in response to small inputs. The phase reduction strategy considers oscillatory dynamics not in terms of the $N$-dimensional state, but rather, in terms of the (one dimensional) oscillation timing yielding a tremendous reduction in dimensionality. Indeed, this approach has been used successfully in many applications where the analysis would otherwise be intractable [4–8].

In situations where the underlying model equations are unknown, $Z(\theta)$ from Eq. (2) can be approximated from data using the direct method [2,9,10]. As a brief description, consider the application of a short pulse of input $u(t) = me_i$ at a known phase $\theta_0$ lasting $t_0$ time units where $e_i \in \mathbb{R}^M$ is the $i$th component of the input, and $m \in \mathbb{R}$ sets the magnitude of the pulse input. By inferring the change in phase $\Delta \theta$ in response to this input, a pointwise estimate of the infinitesimal phase response curve can be estimated according to

$$Z(\theta_0)e_i \approx \frac{\Delta \theta}{m t_0}.\quad (3)$$

By repeating this procedure for multiple choices of $\theta_0$ and for each dimension of the input, $Z(\theta)$ can be inferred by fitting a curve to the resulting data. As a concrete example, consider a conductance-based model neuron of the form (A.1) from Appendix A taking the baseline current $i_b = 8 \mu A/cm^2$ with an additional transmembrane current input $u(t) \in \mathbb{R}$ added to the transmembrane voltage variable. Fig. 1 illustrates an application of the direct method for inputs of varying magnitudes. Panel A shows the response to the example pulse from Panel B which results in the phase shift $\Delta \theta$. Panel C shows the inferred phase response curve with measurements taken according to (3) for inputs of various magnitudes.

Fig. 1 illustrates the primary shortcoming of standard phase reduction — its limited accuracy beyond the weakly perturbed limit. In many practical applications, the applied inputs are far larger than the weakly perturbed limit rendering the phase reduction unusable. A variety of methods have been proposed to better characterize the phase dynamics when larger magnitude inputs are used. For instance, the notion of a residual phase response curve [11–13] has been suggested to characterize how
inputs influence the oscillation timing over multiple periods. Other methods have been developed to understand the phase response to arbitrarily large but slowly \([14,15]\) or rapidly \([16]\) varying inputs. Iterated mapping techniques have also been proposed to consider entrainment of strongly perturbed oscillators to periodic inputs \([17,18]\).

Generally, the phase reduction (2) begins to break down when the magnitude of the applied input is large relative to the size of the smallest magnitude Floquet exponent (which sets the relaxation rate of perturbed solutions back to the limit cycle \([19]\)). Large inputs can drive the state far from the underlying limit cycle so that the infinitesimal phase response curve, which depends on the gradient of the phase with respect to the state evaluated on the limit cycle, no longer provides an accurate approximation of the true response to input. In these situations, amplitude-based effects must be considered in conjunction with the phase dynamics. A variety of phase–amplitude coordinate systems have been suggested in recent years \([14,20–25]\). The amplitude coordinates can be used to characterize how the phase dynamics change as the state is perturbed from the limit cycle \([26]\). Additionally, asymptotic expansions of the phase dynamics in a basis of amplitude coordinates can be helpful for obtaining reduced order equations that are valid for large magnitude inputs \([27,28]\) and for considering phase locking between strongly coupled oscillators \([25,29,30]\). Phase–amplitude reduction techniques have shown great promise for identifying analytically tractable, reduced order models that are valid in the strongly perturbed regime, however, implementation typically requires knowledge of the underlying dynamical equations. Data-driven methods for inference of phase–amplitude-based models are often numerically sensitive to measurement noise, especially when considering asymptotic expansions to high orders of accuracy, limiting implementation in many practical applications.

In this work, data-driven strategies for inference of the necessary terms comprising an adaptive phase–amplitude reduction are considered. Introduced in \([31]\), the adaptive phase–amplitude reduction considers a collection of periodic orbits that results as a given parameter set changes. By actively updating the effective system parameters in order to keep the amplitude coordinates small, a resulting reduced order model can be obtained that is valid far beyond the weakly perturbed limit. In the implementation of this strategy, it is often useful to use isostable coordinates, which are formally defined as level sets of the slowest decaying eigenfunctions of the Koopman operator \([32,33]\), to encode for amplitude-based effects.

Through direct analysis of the equations that underlie the adaptive phase-isostable reduction, this work illustrates that all necessary terms can be inferred by simply measuring a set of infinitesimal phase and isostable response curves (for instance using the direct method or other previously established data-driven strategies \([23,34]\)). In other words, if infinitesimal phase and isostable response curves can be inferred accurately from data, the adaptive phase-isostable reduction can be readily implemented which is generally substantially more accurate than standard phase reduction (2). Furthermore, all necessary terms of the adaptive phase-isostable reduction can also be inferred solely from a series of step function inputs, as opposed to pulsatile inputs typically used for the direct method. By contrast, a previously proposed strategy from \([35]\) required the application of both pulse and step function inputs to infer the necessary terms of the phase-isostable reduction. The analysis provided in this work illustrates that either step function inputs or pulse inputs can be used, and that they ultimately provide redundant information.

The organization of this paper is as follows: Section 2 provides necessary background on the phase, isostable, and adaptive phase-isostable coordinate systems used throughout this work. Section 3 gives a detailed analysis of the terms that comprise the adaptive phase-isostable coordinate reduction, ultimately showing that all necessary terms can be determined with knowledge of the infinitesimal phase response curve along with the infinitesimal isostable response curve and Floquet exponent associated with each isostable coordinate. These relationships suggest an efficient method for inference of the adaptive phase-isostable reduced order equations that require the same data as the well-established direct method. Section 4 provides illustrative examples for a collection of numerical models and Section 5 provides concluding remarks.

2. Background

2.1. Phase coordinates and phase reduction

Consider a general ordinary differential equation of the form (1). Suppose that for a constant choice of \(u = u_0\), Eq. (1) has a stable \(T\)-periodic orbit \(x_{u_0}\). Isochrons \([2,36]\), which represent sets of initial conditions that have the same asymptotic convergence to the limit cycle, can be used to define phase for all states in the basin of attraction of the limit cycle. More precisely, letting \(\theta_1\) be the phase corresponding to an arbitrary initial condition...
\(a(0) \in \mathcal{X}_p\), the \(\theta_i\) isochron, \(\Gamma_{\theta_i}\) corresponds to the set of all \(b(0)\) such that
\[
\lim_{t \to \infty} \|a(t) - b(t)\| = 0,
\]
where \(\| \cdot \|\) can be any vector norm. Using isochrons defined according to Eq. (4) allows for the definition of a phase \(\theta(x, u_0) \in [0, 2\pi)\) for all states \(x \in \mathcal{X}_p\) scaled so that \(d\theta/dt = 2\pi/\omega\) under the flow of (1) when taking \(u = u_0\). Restricting one’s attention to a close neighborhood of the limit cycle, changing variables yields the phase reduction that captures the phase dynamics for inputs that differ from \(u_0\):
\[
d\theta = \frac{\partial \theta}{\partial x} \cdot dx = \frac{\partial \theta}{\partial x} \cdot \left( F(x, u_0) + \frac{\partial F}{\partial u}(u - u_0) + O(\|u - u_0\|^2) \right) = \omega(u_0) + Z(\theta, u_0)(u - u_0) + O(\|u - u_0\|^2)
\]
\[
+ O(\|x - x_0(\theta)\|^2),
\]
where \(Z(\theta, u_0) = \left. \frac{\partial F}{\partial u}(x, u_0) \right|_{u = u_0}\) with all partial derivatives evaluated at \(x = x_0(\theta)\) and \(u = u_0\), the ‘dot’ denotes the dot product, and \(T\) indicates the vector transpose. In the final line of the equation above, \(\frac{\partial F}{\partial u} \cdot x_0(\theta)\) can be used to analyze the \(N\)-dimensional Eq. (1) in terms of a 1-dimensional phase. However, note that the truncation errors grow as both \(u \neq u_0\) and the distance from the limit cycle increase. As such, Eq. (5) is only valid in the weakly perturbed limit \([1,2,4]\), i.e., when \(u - u_0 = O(\epsilon)\) where \(0 < \epsilon \ll 1\). It can still be applied for systems with noninfinitesimal inputs, but it is not guaranteed to match the behavior of the underlying dynamical system (1) in this case. This point is highlighted in Fig. 1.

2.2. Isostable coordinates and phase-isostable reduction

In order to improve the accuracy of the standard phase reduction from Eq. (5), information about the amplitude dynamics must be considered. These amplitude dynamics generally refer to the transient behavior in directions transverse to the limit cycle. A variety of coordinate systems have been proposed \([21-23,25,28]\) that consider the amplitude dynamics. The isostable coordinate framework, which encodes for level sets of the slowest decaying modes of the Koopman operator \([32,37]\), will be considered in this work. When restricting one’s attention to a close neighborhood of the limit cycle, isostable coordinates have a direct relationship with Floquet coordinates. Defining taking \(u = u_0\) and defining \(\Delta x = x - \mathcal{X}_p(\theta)\), to a linear approximation
\[
\Delta x = J\Delta x,
\]
where \(J\) is the Jacobian evaluated at \(x_0(\theta)\). Let \(\Phi\) denote the monodromy matrix, i.e., with the relationship \(\Delta x(T) = \Phi \Delta x(0)\). If \(\Phi\) is diagonalizable, near \(x_0(\theta)\), solutions of (1) follow \([28]\)
\[
x - x_0'(\theta) = \sum_{j=1}^{N-1} \psi_j(\theta, u_0) + O(\psi_1^2) + \cdots + O(\psi_{N-1}^2).
\]
Above, each \(g(\theta, u_0)\) is a Floquet eigenfunction associated with the time varying linear Eq. (6), and \(\psi_1, \ldots, \psi_{N-1}\) are isostable coordinates associated with the periodic orbit \(x_0(\theta)\). For linear time-varying systems, isostable coordinates are often referred to as Floquet coordinates. In contrast to Floquet coordinates, however, isostable coordinates can be defined in the entire basin of attraction of the limit cycle as level sets of the most slowly decaying eigenfunctions of the Koopman operator \([32,38]\); when taking \(u = u_0\) to be constant, these isostable coordinates have dynamics \(\dot{\psi}_j = \kappa_j(u_0)\psi_j\), so that they decay exponentially. Above, \(\kappa_j(u_0)\) is the Floquet exponent associated with the \(j\)th isostable coordinate for the periodic orbit \(x_0(\theta)\). Similar to Eq. (5), through a change of variables one can consider the dynamics of any isostable coordinate for inputs that differ from \(u_0:\)
\[
\frac{d\psi_j}{dt} = \frac{\partial \psi_j}{\partial x} \cdot dx = \frac{\partial \psi_j}{\partial x} \cdot \left( F(x, u) + \frac{\partial F}{\partial u}(u - u_0) + O(\|u - u_0\|^2) \right) = \kappa_j(u_0)\psi_j + f_j(\theta, u_0)(u - u_0) + O(\|u - u_0\|^2)
\]
\[
+ O(\|x - x_0'(\theta)\|^2),
\]
for \(j = 1, \ldots, N - 1\) where \(f_j(\theta, u_0) = \left. \frac{\partial F}{\partial u}(x, u) \right|_{u = u_0}\) with all partial derivatives evaluated at \(x = x_0'(\theta)\) and \(u = u_0\). In the final line above, \(\frac{\partial F}{\partial u} \cdot x_0(\theta)\) because \(\dot{\psi}_j = \kappa_j(u_0)\psi_j\) when \(u = u_0\). Truncating the higher order terms from Eq. (8), the phase reduction from (5) can be augmented with isostable coordinates:
\[
\dot{\theta} = \omega(u_0) + Z(\theta, u_0)(u - u_0),
\]
\[
\dot{\psi}_j = \kappa_j(u_0)\psi_j + f_j(\theta, u_0)(u - u_0),
\]
\[
j = 1, \ldots, N - 1.
\]
In many cases, a subset of the Floquent exponents will have large magnitude real components so that the associated isostable coordinates decay rapidly. Ignoring these rapidly decaying isostable coordinates (assuming they are zero), \(\beta < N - 1\) isostable coordinates can be considered in (9) yielding a reduced order set of equations. Note that while Eq. (9) does take amplitude dynamics into account, it is still only valid in the weakly perturbed limit and generally breaks down when using large magnitude inputs. The adaptive phase-isostable reduction strategy described in the following section can be used to overcome this limitation by considering the dynamics in reference to a continuous family of limit cycles.

2.3. Adaptive phase-isostable reduction

Following the adaptive phase-isostable reduction formulation considered in \([31,35]\), one can consider an additional set of variables \(p \in \mathbb{R}^M\) in the same space as the inputs with dynamics that will be described momentarily. It will be assumed that when \(p\) is held constant, for some set of allowable \(p\), the differential equation \(\dot{x} = F(x, p)\) has a stable periodic orbit \(x_0\). For each \(x_0\), a corresponding phase \(\theta(x, p)\) and set of isostable coordinates \(\psi_1(x, p), \ldots, \psi_N(x, p)\) can be defined. As discussed in \([31]\), the phase coordinates associated with each \(x_0(\theta)\) are unique up to a constant shift. These shifts can be explicitly defined, for instance, using a Poincaré section to correspond to a level set of the phases between limit cycles.

To proceed, one can rewrite Eq. (1) as
\[
\dot{x} = F(x, u) + U(x, u, p),
\]
where
\[
U(x, u, p) = F(x, u) - F(x, p)
\]
\[
= \frac{\partial F}{\partial u}(u - p) + O(\|u - p\|^2),
\]
where all partial derivatives are evaluated at \(x\) taking \(u = p\). In Eq. (10), \(F(x, p)\) represents the nominal behavior of the system (1) when taking \(u = p\). The term \(U(x, u, p)\) can be thought of as an effective input. Using this formulation, let \(p\) be nonstatic. Changing variables to work in a basis of phase and isostable coordinates yields
\[
\frac{d}{dt} \theta(x, p) = \frac{\partial \theta}{\partial x} \cdot dx + \frac{\partial \theta}{\partial p} \cdot dp,
\]
where $\frac{dp}{dt} = G_p(p, \theta, \psi_1, \ldots, \psi_\beta)$.

Above, $D(\theta, p) \in \mathbb{R}^M$ with the $i$th element given by $-\frac{\partial x_i}{\partial \theta} \cdot \frac{\partial x_i}{\partial p}$ where $\frac{\partial x_i}{\partial \theta} \bigg|_{\theta_o, p_o}$ is defined by $x_i' = \lim_{\theta \to \theta_o} (x_i + \varepsilon\partial x_i/\partial \theta_0) / \varepsilon$ where $x_i$ is evaluated in reference to $x_i'$. Each $\gamma_i$ is the $i$th component of the standard unit basis. Likewise $Q_i(\theta, p) \in \mathbb{C}^M$ with the $i$th element given by $-\frac{\partial x_i}{\partial \theta} \cdot \frac{\partial x_i}{\partial p}$ where $x_i$ is evaluated in reference to $x_i'$. Note that each $\psi_j$ can potentially take complex values so that $Q_i$ can also take complex values. When using Eq. (13) to represent the phase and isostable coordinate dynamics, Eq. (7) can be used to approximate the state. More details about this derivation are provided in [31].

The primary sources of error in the standard phase reduction (5) are associated with large deviations from the limit cycle and the application of large inputs relative to the baseline value. The variable $\psi$ in Eq. (13) adds additional degrees of freedom that can mitigate these errors. For instance, provided $G_p$ from Eq. (13) can be found so that $\psi_1(x, \theta), \ldots, \psi_j(x, \theta)$ remain small, $x - x_i'$ will remain small, thereby limiting truncation errors that are associated with being perturbed far from an underlying limit cycle. General heuristics for the design of $G_p$ are discussed in [31]. In a specific situation where only one isostable coordinate is necessary, $p \in \mathbb{R}^1$, and $Q_i(\theta, p)$ is bounded away from 0 for all $\theta$ and all allowable $p$, one can take $G_p = -l_i(\theta, p) Q_i(\theta, p)(u - p)$.

so that the isostable coordinate dynamics follow $\dot{\psi}_1 = \kappa_1(p) \psi_1$ and hence can be ignored. This yields the resulting 2-dimensional reduction $\dot{\theta} = \omega(p) + Z(\theta, p)(u - p) - D(\theta, p) I_1(\theta, p) Q_1(\theta, p)(u - p)$.

\[ \dot{p} = -\frac{l_1(\theta, p)}{Q_1(\theta, p)}(u - p). \]

More generally, one can also use $G_p = -v \psi_1 Q_i(\theta, p)$.

where $v > 0$ so that the isostable coordinate dynamics become $\dot{\psi}_1 = \kappa_1(p) - v Q_i(\theta, p) \dot{\theta} + I_1(\theta, p)(u - p)$. In this case, $G_p$ effectively increases the decay rate of transients. Importantly, as explained in [31], the adaptive phase-isostable reduction strategy can be used to analyze the dynamics of (1) in a reduced order setting without placing limitations on the magnitude of the allowable inputs.

2.4. Operational phase coordinates

In much of this work, numerical approximations of infinitesimal phase and isostable response curves will be obtained using strategies similar to the direct method [29,10]. These strategies are implemented by applying a small input at a known initial phase and measuring the resulting changes in the phase and isostable coordinates. The asymptotic phase is relatively straightforward to measure by allowing the state to relax back to the limit cycle after the perturbation. Inference of isostable coordinates (that capture transient behavior) is more difficult necessitating the consideration of the operational phase described in [23]. This coordinate system distinguishes between asymptotic phase, $\theta$, defined according to the isochrons (4) and an operational phase, $\theta^*$, defined so that $\theta^*(x) = 0$ every time the state crosses a predefined Poincaré section.

The distinction between operational and asymptotic phase coordinates is necessary because it is generally not possible to precisely measure the asymptotic phase $\theta(x, p)$ for any $x \neq x_i'$. By contrast, the precise timing of the crossing of the $\theta^* = 0$ level set (i.e., Poincaré section) can always be determined once per cycle. As discussed in Appendix D, a number of relations are available that connect the operational phase to the asymptotic phase. For instance, the reduced order dynamics can be accurately captured using a single isostable coordinate $\psi_1$, to linear orders of accuracy $t^* = \tau_1 + \psi_1(t_1) \kappa_1$, \[ g(0, p) \cdot n = \frac{\psi_1'\gamma_0 - n}{\kappa_1(p)}, \]

where $\gamma \equiv d/d\theta$ and $n$ is a unit vector normal to the Poincaré section used to define $\theta^* = 0$. Finally, as discussed in [23], features of the isostable coordinate dynamics can be inferred by considering an initial condition that has been perturbed from its limit cycle at $x_i'$. For instance, in the case where there is a single isostable coordinate, the associated Floquet exponent $\kappa_1$ can be obtained according to $\kappa_1 = \log(t_{n+1} - t_n)/T(p)$.

where $\tau_1 = t_{n+1} - t_n - T(p)$. Note that when noise is present, each $\tau_1$ used in the estimate of (19) must be taken soon after the initial perturbation; otherwise the measurement of $\tau_{n+1}/\tau_n$ in Eq. (19) will be highly susceptible to the noise. Relations (17)-(19) will be used in conjunction with the proposed model identification strategy as described in Section 4. These equations are obtained directly from results presented in [23] as shown in Appendix D.

3. Inference of adaptive phase-isostable reduced models from first order accurate terms

The direct method [29,10] and related approaches [23,34] have been previously proposed to infer $Z(\theta, p)$ and $I_1(\theta, p)$ from (13) for a given $x_i'$ by considering the phase response to pulsatile inputs. Ref. [35] suggests a similar strategy for finding $D(\theta, p)$ and each $Q_i(\theta, p)$ using step function inputs instead of pulses. Alternatively, the derivation presented below illustrates that the additional terms of the adaptive reduction $D(\theta, p)$ and $Q_i(\theta, p)$ are directly related to $Z(\theta, p)$ and $I_1(\theta, p)$, respectively. These relations are obtained by first considering the shift of the periodic orbit in response to a static perturbation using an isostable coordinate basis.
3.1. Response to static perturbations using an isostable coordinate basis

To begin, consider a periodic orbit $x_p^o(\theta)$ of the general dynamical system (1) that results when holding $u$ constant at the value $p$. Considering the solution in a basis of phase and isostable coordinates with dynamics governed by Eq. (9), the solution $\theta(t)$ and $\psi_j(t)$ in response to the static shift in input $\Delta u(t) = p + \epsilon c_1$ can be expanded in powers of $\epsilon$ where

$$\theta(t) = \theta_0(t) + \epsilon \theta_1(t) + \epsilon^2 \theta_2(t) + \ldots, \quad (20)$$

$$\psi_j(t) = \psi_{j,0}(t) + \epsilon \psi_{j,1}(t) + \epsilon^2 \psi_{j,2}(t) + \ldots, \quad (21)$$

for $j = 1, \ldots, N - 1$. First considering the solution $\theta(t)$, substituting Eq. (20) in to Eq. (9) yields

$$\dot{\theta}_0 + \epsilon \dot{\theta}_1 + \epsilon^2 \dot{\theta}_2 + \ldots = \omega + \epsilon Z'(\theta_0(t)) + \epsilon \theta_1(t) + \epsilon^2 \theta_2(t) + \ldots, \quad (22)$$

where $Z'$ is the $i$th component of $Z$. Note that the dependence of $\omega$ on $p$ has been suppressed for convenience of notation. Collecting $O(1)$ terms, $\dot{\theta}_0 = \omega$ so that $\theta_0(\theta(t)) = \theta(0) + \omega t$. For simplicity of exposition, let $\theta(t = 0) = 0$. Continuing with the $O(\epsilon)$ terms

$$\dot{\theta}_1 = Z'(\theta_0(t)),$$

so that

$$\theta_1(t) = \int_0^t Z'(\omega s, p) \, ds + c_1, \quad (24)$$

where $c_1$ is a constant to be determined momentarily. Next considering the isostable coordinates, substituting Eq. (21) into Eq. (9) yields

$$\dot{\psi}_{j,0} + \epsilon \dot{\psi}_{j,1} + \ldots = -\kappa_j(\psi_{j,0}(t) + \epsilon \psi_{j,1}(t) + \ldots) + \epsilon \kappa_j \theta_0(t), \quad (25)$$

for $j = 1, \ldots, N - 1$ where $l_j$ is the $j$th component of $I_j$. Once again, the dependence of $\kappa_j$ on $p$ has been suppressed for convenience of notation. Collecting $O(1)$ terms, $\dot{\psi}_{j,0} = \kappa_j \psi_{j,0}$ so that $\psi_{j,0}(t) = \psi_{j,0}(0) \exp(\kappa_j t)$ and lim$_{t \to \infty} \psi_j(t) = 0$. Continuing with the $O(\epsilon)$ terms

$$\dot{\psi}_{j,1} = \kappa_j \psi_{j,1} + I_{l_j}(\omega s, p),$$

where $c_{l_j,0}$ is the Fourier series coefficients of $I_{l_j}(\omega s, p)$. After all transients have decayed Eq. (26) admits periodic solutions of the form

$$\psi_{j,1,\text{st}}(t) = \frac{b^i_{l_j,0}}{\kappa_j} - \sum_{m=1}^{\infty} \frac{b^i_{m,i} \kappa_j}{\kappa_j^2 + \omega^2 m^2} \sin(m \omega t) + \frac{b^i_{m,i} \kappa_j^2 \omega}{\kappa_j^2 + \omega^2 m^2} \cos(m \omega t) + \frac{b^i_{0,i} \kappa_j}{\kappa_j^2 + \omega^2 m^2} \sin(m \omega t) + \frac{b^i_{0,i} \kappa_j^2 \omega}{\kappa_j^2 + \omega^2 m^2} \cos(m \omega t). \quad (27)$$

With the above results in mind, next consider the expansions for $\theta(t)$ and $\psi_j(t)$ from Eqs. (20) and (21). For Eq. (20), to leading order $\epsilon$

$$\theta(t) = \omega t + \epsilon \left[ \int_0^t Z'(\omega s, p) \, ds + c_1 \right] + O(\epsilon^2). \quad (28)$$

After all transients have decayed, Eq. (21) becomes

$$\psi_{j,1,\text{st}}(t) = \psi_{j,1,\text{st}}(0) + O(\epsilon^2). \quad (29)$$

Towards simplifying Eqs. (28) and (29) to find a periodic solution, let $G(\omega s, p) = Z'(\omega s, p) - \tilde{Z}(p)$, where $\tilde{Z}(p) = \frac{1}{2} \int_0^p Z'(\omega s, p) \, ds$. Eq. (28) can be rewritten as

$$\theta(t) = \omega t + \epsilon \left( H'(\omega t, p) + c_1 \right) + O(\epsilon^2), \quad (30)$$

where $H'(\omega t, p) = \int_0^t G'(\omega t, p) \, ds$ and $\omega^2 = \omega + \epsilon \tilde{Z}'(p)$. Considering Eq. (7) which relates the phase and isostable coordinates to the state, substituting Eqs. (29) and (30) for the phase and isostable coordinates (which gives the solution after all transients have decayed), and neglecting $O(\epsilon^2)$ terms, the periodic solution resulting from the static shift in input is

$$x_{p+\epsilon c_1}^r(o_d A t) = x_p^r(o_d A t) + \epsilon H'(\omega t, p) + c_1 + \sum_{j=1}^{N-1} [ \epsilon g_j(\theta) \psi_{j,1,\text{st}}(\omega t) ], \quad (31)$$

Taylor expanding $x_p^r$ and each $g_j$ and $\psi_{j,1,\text{st}}$ about $o_d A t$ from the above equation and neglecting $O(\epsilon^2)$ terms ultimately yields

$$x_{p+\epsilon c_1}^r(\theta) = x_p^r(\theta) + \epsilon x_p^r(\theta) [H'(\omega, p) + c_1] + \sum_{j=1}^{N-1} [ \epsilon g_j(\theta) \psi_{j,1,\text{st}}(\theta) ], \quad (32)$$

where $\dot{\theta} = \frac{d\theta}{dt}$ and $\hat{\theta} = \omega_d t$ leading order $\epsilon$ along the new periodic orbit. The constant shift $c_1$ is related to the fact that each $x_p^r$ is unique up to a constant phase shift.

Eq. (32) gives an $O(\epsilon)$ accurate approximation for the shift in the periodic orbit in response to a static change in parameters. Note that Eq. (32) shares similarities with results from [39]. The advantage here is that working in an isostable coordinate basis allows for substantial simplification when considering the terms of the adaptive phase-isostable reduction from (12) as illustrated in Sections 3.2 and 3.3.

3.2. Inferring terms of the adaptive phase-isostable reduction with only pulse inputs

Considering Eq. (32) and the definition of $\frac{\partial x_p^r}{\partial \theta}$ given under Eq. (13), one can write

$$\frac{\partial x_p^r}{\partial \theta} = x_p^r(\theta) + \sum_{j=1}^{N-1} \left[ g_j(\theta) \psi_{j,1,\text{st}}(\theta) \right], \quad (33)$$

where $\theta$ is the asymptotic phase. Toward identification of the terms $D(\theta, p)$ and $Q_1(\theta, p)$ from the adaptive phase-isostable reduction (13), as discussed in [27], $\frac{\partial x_p^r}{\partial \theta} = 1$ and $\frac{\partial x_p^r}{\partial \theta} \cdot g_j(\theta, p) = 0$ for all $j$, with all partial derivatives and functions evaluated at $\theta$ on the $x_p^r$ limit cycle. With this in mind, $D(\theta, p)$ can be written as

$$D(\theta, p) = -\frac{\partial x_p^r}{\partial p_i} \cdot \frac{\partial \theta}{\partial x} = -H'(\omega, p) - c_1. \quad (34)$$

Likewise, as discussed in [27], $\frac{\partial x_p^r}{\partial \theta} \cdot g_j(\theta, p) = 1$ if $j = n$ and 0 otherwise and $\frac{\partial x_p^r}{\partial \theta} \cdot g_j(\theta, p) = 0$. Using this result in conjunction with Eq. (33) one finds

$$Q_1(\theta, p) = -\frac{\partial x_p^r}{\partial p_i} \cdot \frac{\partial \psi_{j,1,\text{st}}}{\partial x} = -\psi_{j,1,\text{st}}(\theta). \quad (35)$$

Recall from Eq. (27) that $\psi_{j,1,\text{st}}(\theta)$ is fully specified by the Fourier coefficients of $I_1(\omega s, p)$, and $c_1$. Likewise, $H'(\omega, p)$ is fully specified by $Z'(\omega, p)$. As such, with the exception of $c_1$, the terms that comprise (34) and (35) are uniquely determined with knowledge of the first order accurate phase-isostable dynamics from (9). In determining the value of the constant $c_1$, recall that each $x_p^r(\theta)$ is unique up to a constant phase shift in the implementation of
the adaptive phase–amplitude reduction (13). To disambiguate this constant phase shift, for instance, one can define a Poincaré section S as an n−1 dimensional hyperplane and take  in S to correspond to the intersection of S and  for all p. With this definition,  for all p. With this definition,  evaluated at  is constant, where n is a unit normal vector to S. Using this definition, starting with (32), taking the dot product with n on both sides and rearranging one finds

\[ c_i = -\sum_{j=1}^{n-1} \left( (g_i(0, p) \cdot n) \psi_{i,ss}(0) \right) - (x'_0(0) \cdot n) H'(0, p) \].

(36)

Above, H(0, p) and each  can be determined directly from the infinitesimal phase and isostable response curves. In applications where only one isostable coordinate is required,  can be determined according to Eq. (18). As such, by inferring the first order accurate terms for a collection of periodic orbits that comprise (7) and (9), the necessary terms of the adaptive phase-amplitude reduction from (13) are uniquely determined. Note that each  from Eq. (34) does not depend on the isostable coordinates and each  from Eq. (35) only depends on the dynamics associated with theψs isostable coordinate. As such, model order reduction that ignores the most rapidly decaying isostable coordinates is still possible when computing D(θ, p) and each necessary Qs(θ, p) using a reduced order isostable coordinate basis.

3.3. Inferring terms of the adaptive phase–isostable reduction with only step function inputs

As discussed in (35), it is possible to infer the terms D(θ, p) and each Qs(θ, p) from Eq. (13) using step function inputs. For instance, suppose u(t) = p + εh(t), where h is the Heaviside step function and ε is the ith component of the standard unit basis. Using this step function input and letting  with associated  respectively, one can obtain pointwise estimate of D'(θ(t, p), p) = δθ/ε and Q's(θ(t, p), p)/ε. This process can be repeated for multiple values of θ(t) and subsequently for different initial values of p allowing D(θ, p) and each Qs(θ, p) to be fit to the resulting data.

Considering the results from Sections 3.1 and 3.2, it is also possible to apply a series of step function inputs to infer D(θ, p) along with each Qs(θ, p) and subsequently compute Z(θ, p) and each Is(θ, p). Starting with the inference of the infinitesimal phase response curve, Z(θ, p), consider Eq. (34) after substituting θ = t along the limit cycle:

\[ D'(\omega t, p) = -H'(\omega t) - c_i, \]

\[ = \int_0^t -Z'(\omega s, p) + \ddot{Z}(p)ds - c_i. \]

(37)

Taking the time derivative and rearranging yields

\[ Z'(\omega t, p) = -\frac{d}{dt} D'(\omega t, p) + \ddot{Z}(p). \]

(38)

As discussed elsewhere [39], to leading order the shift in frequency is directly related to the average value of the infinitesimal phase response curve, i.e., \( \ddot{Z}(p) \approx \omega(p + \epsilon \delta) - \omega(p)/\epsilon \). Thus, an estimate of \( Z(\theta, p) \) from (13) can be obtained with knowledge of both D(θ, p) and the shift in the natural frequency resulting from a static shift in the parameters.

Next, for inference of the infinitesimal isostable response curves, letting \( Q_s(\theta, p) = b^{\dagger}_{0,j} + \sum_{m=1}^{\infty} \left[ a_{mj}^i \sin(\nu \theta) + b_{mj}^i \cos(\nu \theta) \right] \) where \( a_{mj}^i \) and \( b_{mj}^i \) are Fourier coefficients of \( Q_s(\theta, p) \), and considering Eq. (35), after algebraic manipulation one finds

\[ b_{0,j}^i + \sum_{m=1}^{\infty} \left[ a_{mj}^i \sin(\nu \theta) + b_{mj}^i \cos(\nu \theta) \right] \]

\[ = -\psi_{ss,j}, \]

\[ = b_{0,j}^i + \sum_{m=1}^{\infty} \left[ a_{mj}^i \frac{\nu \sin(\nu \theta)}{\nu^2 - \omega^2m^2} + \frac{a_{mj}^i \nu \omega \cos(\nu \theta)}{\nu^2 - \omega^2m^2} \right] \]

\[ - b_{mj}^i \frac{\nu \sin(\nu \theta)}{\nu^2 - \omega^2m^2} + \frac{b_{mj}^i \nu \omega \cos(\nu \theta)}{\nu^2 - \omega^2m^2}. \]

(39)

where the final line is obtained using Eq. (27). Subsequently, equating Fourier coefficients in Eq. (39) yields

\[ b_{0,j}^i = b_{0,j}^i k_j, \]

\[ \begin{bmatrix} a_{mj}^i \\ b_{mj}^i \end{bmatrix} = \begin{bmatrix} k & m \omega \\ -m \omega & k \end{bmatrix} \begin{bmatrix} d_{mj}^i \\ b_{mj}^i \end{bmatrix}. \]

(40)

Recall here that \( d_k \) and \( b_k \) are the Fourier series coefficients of \( I_i(\theta) \) as defined below Eq. (26). Eq. (40) provides a direct relationship between the Fourier coefficients of \( Q_s(\theta, p) \) and the Fourier coefficients of \( I_i(\theta, p) \) from Eq. (13).

3.4. Summary of analytical results

In summary, the adaptive phase–isostable reduction from Eq. (13) is comprised of terms from the first order accurate phase–isostable reduction from Eq. (9) (i.e., \( \omega(p), Z(\theta, p) \), and both \( I_i(\theta, p) \) and \( k_j(p) \) for each nontruncated isostable coordinate) along with the terms \( D(\theta, p) \) and \( Q_s(\theta, p) \) for each nontruncated isostable coordinate. As illustrated in the previous sections, the terms \( D(\theta, p) \) and \( Q_s(\theta, p) \) can be computed explicitly with knowledge of \( \omega(p), Z(\theta, p) \), \( I_i(\theta, p) \), and \( k_j(p) \). Additionally, \( \omega(p), Z(\theta, p) \), and \( I_i(\theta, p) \) can be computed explicitly with knowledge of \( \omega(p), D(\theta, p), Q_s(\theta, p) \), and \( k_j(p) \). As such, a direct method approach for the computation of these necessary terms can be implemented in one of the following two ways:

1. Using pulsatile inputs, the resulting changes to the phase and isostable coordinates can be inferred using strategies similar to the direct method [23, 34]. Subsequently, \( Z(\theta, p) \) and each I_i(\theta, p) with associated k_j(p) can be determined for a collection of periodic orbits x^i. The remaining terms \( D(\theta, p) \) and each Qs(\theta, p) can be determined according to Eqs. (34) and (35), respectively.

2. Using step function inputs the resulting changes to the phase and isostable coordinates can be inferred using strategies similar to the direct method [23, 34]. Subsequently, \( D(\theta, p) \) and each Qs(\theta, p) with associated k_j(p) can be determined for a collection of periodic orbits x^i. The remaining terms \( Z(\theta, p) \) and each I_i(\theta, p) can be determined according to Eqs. (40) and (35), respectively.

When using either strategy 1 or 2 above, \( \omega(p) \) can be inferred from the unperturbed natural period. Fig. 2 provides a schematic reasoning for the theoretical results from Section 3.

The computational results to follow in Section 4 illustrate both of these model identification approaches for a collection of numerical models. Results from Sections 4.2 and 4.3 use strategies 1 and 2, respectively, but either strategy could be implemented in both of these applications.
4. Numerical illustration

4.1. Illustration in a simple dynamical model with an analytical solution

The derived relationships (34), (35), (38), and (40) are illustrated for a simple dynamical model

\[
\begin{align*}
\dot{x} &= \pm \xi_0(\theta, p_1^2(t) - r^2) = y(t) + \eta(t), \\
\dot{y} &= \pm \xi_0(\theta, p_1^2(t) - r^2) + x(t) + \eta(t).
\end{align*}
\]

Here, \( x \) and \( y \) are state variables representing Cartesian coordinates, \( r = x^2 + y^2 \), \( \xi \) and \( \eta \) are model parameters, and \( \mu \) is a bifurcation parameter. Stable oscillations emerge resulting from a Hopf bifurcation when \( \mu > 0 \). Here, \( \mu \) will also be considered as a control input. For constant values of \( \mu \), Eq. (41) admits stable periodic solutions \( x_0(t) = [\mu \cos(t), \mu \sin(t)]^T \). Because Eq. (41) is two-dimensional, each periodic orbit has one isostable (amplitude) coordinate. Analytic solutions for the reduced order equations associated with Eq. (41) can be obtained as described below.

To begin, the gradient of the phase can be obtained by finding appropriately normalized periodic solutions of the adjoint Eq. (C.1). For this model, these solutions are \( \{d\phi/dx, d\phi/dy\} = [\eta \cos(t)/\sigma \mu - \sin(t)/\mu, \eta \sin(t)/\sigma \mu + \cos(t)/\mu] \). Additionally, \( \frac{\delta}{\delta \mu} = [2\mu^2(\eta \sin(t) + \sigma \cos(t)), 2\mu^2(\sigma \sin(t) - \eta \cos(t))] \). Directly computing the infinitesimal phase response curve using the relation provided just below Eq. (5),

\[
Z(\theta, \mu) = \left[ \frac{\partial \phi}{\partial x}, \frac{\partial \phi}{\partial y} \right] \frac{dF}{d\mu} = 0.
\]

Furthermore, \( \frac{\delta}{\delta \mu} = [\cos(t), \sin(t)]^T \) so that using the relation provided just below Eq. (13),

\[
D(\theta, \mu) = -\frac{\partial^2 \phi}{\partial \mu} \left[ \frac{\partial \phi}{\partial x}, \frac{\partial \phi}{\partial y} \right] = -\frac{\eta}{\mu^2}.
\]

Notice that Eq. (38) holds for \( Z(\theta, \mu) \) and \( D(\theta, \mu) \) above, i.e., that \( \frac{d\theta}{dt} = 0 \) so that \( Z(\theta, \mu) \) must be a constant. Because the frequency does not depend on \( \mu \), the right hand side of Eq. (38) is zero yielding the correct infinitesimal phase response curve. Similarly, it is straightforward to verify that Eq. (34) holds here.

Continuing for the isostable coordinate dynamics, in radial coordinates, Eq. (41) becomes

\[
\begin{align*}
\dot{r} &= \sigma(\mu t^2 - r^2), \\
\dot{v} &= 1 + \eta(t^2 - \mu t^2),
\end{align*}
\]

(44)

valid for \( \mu > 0 \) where \( \alpha_1 \) is a constant determined by the initial condition. In the limit as time approaches infinity, Eq. (45) is well approximated by \( r(t) \approx \alpha_2 \exp(-2\mu^2 t^) \) where \( \alpha_2 \) is a constant that depends on initial conditions. As such, the Floquet exponent is \( \kappa_1(\mu) = -2\mu^2 \sigma \). Eq. (C.2) admits periodic solutions of the form \( \{d\psi_1/dx, d\psi_1/dy\} = [\cos(t), \sin(t)] \). Taking \( \mu(t) \) as the input, one finds

\[
\begin{align*}
I_1(\theta, \mu) &= \left[ \frac{d\psi_1}{dx}, \frac{d\psi_1}{dy} \right] \frac{dF}{d\mu} = 2\mu^2 \sigma, \\
Q_1(\theta, \mu) &= \left[ \frac{\partial^2 \psi_1}{\partial \mu^2} \right] \left[ \frac{d\psi_1}{dx}, \frac{d\psi_1}{dy} \right] = -1,
\end{align*}
\]

(46)

where \( I_1 \) is defined below Eq. (9) and \( \kappa_1(\mu) \) is defined below Eq. (13). Once again, the relationship (40) mandates that \( I_1(\theta, \mu) = -\kappa_1 = 2\mu^2 \sigma \) as expected. It is also straightforward to verify that Eq. (35) holds. With knowledge of the terms of the adaptive reduction, one can use (15) to represent Eq. (41) as

\[
\begin{align*}
\dot{\theta} &= 1 + 2\eta(p - \mu(t)), \\
\dot{p} &= 2p^\theta \sigma(\mu(t) - p).
\end{align*}
\]

(47)

Notice that Eq. (47) shares similarities with the polar form of the model Equations from (44), but is not identical.

Fig. 3 gives a comparison between the adaptive model from (47) and the phase-isostable-based model of the form (9). Note
4.2. Illustration in a noisy model of a tonically firing neuron

Here, the proposed model identification algorithm is illustrated using the Wang-Buzsáki model of neural spiking behavior [40],

\[
C \dot{V} = -g_{Na} m^3 h (V - E_{Na}) - g_K n^4 (V - E_K) - g_L (V - E_L)
\]

\[
\dot{h} = \gamma [\alpha_h(V)(1 - h) - \beta_h(V)h],
\]

\[
\dot{n} = \gamma [\alpha_n(V)(1 - n) - \beta_n(V)n],
\]

\[
\dot{w} = \omega (1.5/(1 + \exp(-b - V/k)) - w).
\]  

Above, \( V \) represents the transmembrane voltage, \( h \) and \( n \) are gating variables, and \( i_b(t) \) is a baseline current (in \( \mu A/cm^2 \)), and \( \sqrt{\frac{2\xi}{C}} \chi(t) \) is zero-mean Gaussian noise with intensity \( \xi = 0.2 \).

Here, the dynamical equations have been augmented with an adaptation current \([41], i_a = g_w w (V - E_K) \) where \( g_w = 2 mS/cm^2 \) and \( E_K = -90 mV \). A full description of the model equations and parameters is given in Appendix A.

The baseline current \( i_b \) is taken to be an input. Writing Eq. (48) in the same form as Eq. (10) yields

\[
U_i(x, i_b(t), p) = \left[ (i_b(t) - p) \ 0 \ 0 \ 0 \right]^T, \text{ where } x = [V \ h \ n \ w]^T.
\]

For constant values of \( i_b \) in \([4, 20]\) \( \mu A/cm^2 \), the neuron is in a periodically firing regime with period \( T \in [3.8, 14.3] \). Here, larger baseline currents correspond to shorter unperturbed periods. For any \( x \) on the periodic orbit \( x_{b}^p \), an action potential is defined to occur at the moment the transmembrane potential crosses 0 with a positive slope. Here, the crossing of this \( V = 0 \) Poincaré section is taken to correspond to the \( \theta = 0 \) level set for all periodic orbits.

4.2.1. Direct method inference of infinitesimal phase and isostable response curves using pulsatile inputs

A model of the form (13) with a single isostable coordinate will be inferred from noisy simulations of (48) using a slight modification of the direct method [9,10] to compute \( Z(\theta, i_b), \kappa(i_b), \) and \( l_1(\theta, i_b) \) associated with \( x_b^s \) for \( i_b \in \left[ 4, 20 \right] \). This approach employs an operational phase–amplitude coordinate system [23].

For a given choice of \( \theta_0 \) and \( i_0 \) pulse inputs of the form

\[
i_0(t) = \begin{cases} 
\ i_0 + \Delta i, \quad & t_s + \theta_0 \frac{T(i_b)}{2\pi} < t < t_s + \theta_0 \frac{T(i_b)}{2\pi} + \Delta t \ \\
\ i_0, \quad & \text{otherwise}, 
\end{cases}
\]  

are applied where \( t_s \) denotes the timing of the most recent action potential, \( \Delta i \) corresponds to the magnitude of a pulse input lasting \( \Delta t \) time units, and \( t_{\text{max}} \) sets the duration of series of pulse trains. Fig. 4 highlights the pulsing protocol mandated by Eq. (49).

Assume that \( t_{\text{max}} \) is chosen to be large enough so that the effect of initial conditions dies out, during the application of pulse inputs \( k \) th interspike interval will be

\[
t^*_k - t^*_k = t_{k+1} - t_k - \frac{\alpha_1}{\omega_k} (\psi_1(t_{k+1}) - \psi_1(t_k)) 
= t_{k+1} - t_k.
\]  

Derivation of the above equation uses Eq. (17) along with the fact that \( \psi_1(t_{k+1}) = \psi_1(t_k) \) in steady state. As such, the \( k \) th interspike interval corresponds to the time between subsequent crossings of the \( \theta = 0 \) isochron. Note that in the above equation, the dependence of \( \kappa_1 \) and \( \omega \) on \( i_0 \) is suppressed for convenience of notation. Considering the form of the reduced order Eq. (13) with \( p = i_0 \) held constant, in steady state, the change in phase resulting from the pulse occurring on each cycle is well approximated by

\[ \Delta \theta \approx Z(\theta_0, i_0) \Delta t \Delta \theta. \]

The difference in the interspike interval will be

\[
t^*_k - t^*_k = t_{k+1} - t_k = T(i_b)(1 - \Delta \theta/2\pi) \]

so that a pointwise estimation of the infinitesimal phase response curve can be obtained according to

\[
Z(\theta_0, i_0) \approx \frac{2\pi}{T(i_0) \Delta t \Delta \theta} (dt^* - T(i_0)),
\]
where \( dt^* = \frac{1}{N} \sum_{i=1}^{N} (t^*_{i+1} - t^*_i) \) is taken as the average of the \( N \) preceding interspike intervals in order to average out the effects of noise. To infer pointwise estimates of the infinitesimal isostable response curve, one can consider the decay back to the limit cycle after the pulses stop. Letting \( t_p \) be the moment immediately after the final pulse is applied, as discussed in [23], each subsequent action potential occurs when

\[
2\pi n = \theta^*(t_p) + \omega(t^*_{p,n} - t_p) + \frac{\psi(t_p)}{\kappa_1} \exp(\kappa_1(t^*_{p,n} - t_p) - 1),
\]

is satisfied, where \( t^*_{p,n} \) is the \( n \)th spike after the final pulse input is applied. Comparing the first and the \( n \)th spikes after the final pulse is applied, one finds

\[
\psi_1(t_p) = \frac{\kappa_1(2\pi(n-1) - \omega(t^*_{p,n} - t^*_{1,n}))}{\exp(\kappa_1(t^*_{p,n} - t_p)) - \exp(\kappa_1(t^*_{p,n} - t_p))}.
\]

Once again, considering the isostable coordinate dynamics of the reduced order Eq. (13) with \( \rho = \dot{i}_0 \) held constant, immediately after each pulse is applied,

\[
\psi_1(t_p) \approx I_1(\theta_p, \dot{i}_0) \Delta t + \psi_1(t_p) \exp(\kappa_1 dt^*),
\]

where approximate equality comes from the fact that the effect of initial transients have died out so that the isostable coordinate immediately preceding the previous pulse is identical to \( \psi_1(t_p) \) and \( dt^* \) is used in order to average out the effects of noise. Rearranging yields pointwise estimates of the infinitesimal isostable response curve

\[
I_1(\theta_p, \dot{i}_0) \approx \frac{\psi_1(t_p)(1 - \exp(\kappa_1 dt^*))}{\Delta t},
\]

where \( \psi_1(t_p) \) is computed according to from Eq. (53). Finally, as discussed in [23] using the same time series data, a representative estimate of \( \kappa_i(p) \) can be obtained according to (19).

4.2.2. Reduced order model validation

Infinitesimal phase response curves are approximated according to the strategy described in Section 4.2.1 taking \( \dot{i}_0 \in [4, 20] \mu A/cm^2 \). Results are shown in Fig. 5 for the curves obtained for the \( x^{i}_{v} \) periodic orbit taking \( \dot{i}_0 = 12 \mu A/cm^2 \). Open circles in panels A and D are obtained according to Eqs. (51) and (55), respectively. Infinitesimal phase and isostable response curves are fit using a Fourier series basis and shown as solid lines. Panels B and E compare the Fourier series fits to the data (thick black lines) to the true infinitesimal phase and isostable response curves (thin black lines) obtained from periodic solutions to Eqs. (C.1) and (C.2), respectively. Panels C and F show \( D(\theta) \) and \( Q(\theta) \) inferred from relations given in Section 3.2 using the infinitesimal phase and isostable response curves obtained from the direct method (thick black lines). These are compared to the true values (thin black lines).
vertical offset between the true curves and the curves obtained from noisy data. Results shown in Fig. 5 are representative of the results obtained when taking different constant values for periodic orbits $\chi^p$ for different choices of $i_b$. Despite the fact that the direct method does not perfectly identify the terms comprising the adaptive phase-isostable reduced order Eqs. (13), the resulting model still accurately captures the response to general inputs as shown in the results below.

Outputs from both the adaptive phase-isostable reduction from Eq. (13) and the phase-only reduction of the form (5) are compared to full model simulations using sinusoidal inputs with different frequencies. Results are shown in Fig. 6. For the adaptive reduction, the update rule $\dot{\psi} = \mathcal{G}(i_b, \theta, \psi_{t1}) = -1000 Q_1(\theta, i_b) \psi_{t1}$ and is of the same form as (16). The output $V(t)$ is taken to be $V^\#(\theta(t))$, that is, the voltage corresponding to $\theta(t)$ on the $x^\#_T$ periodic orbit. The phase-only reduction uses infinitesimal phase response curves associated with the periodic orbit that results when taking $i_b = 10 \ \mu A/cm^2$. Panels A, B, and C, show the transmembrane voltage output for each model and panels D, E, and F show the resulting difference between the phase from the reduced order model, $\theta_{\text{red}}$, and the full order simulation, $\theta_{\text{act}}$, in response to the inputs $i_b(t)$ in panels G, H, and I, respectively. Note that panels D, E, and F, only show discrete measurements because the phase of the full order simulations can be measured once per cycle when the neuron fires an action potential (thereby crossing the $\theta = 0$ level set). Also, while the reduced order models were inferred using noisy data, the simulations from Fig. 6 take the noise intensity to be $\zeta = 0$ in order to compare the accuracy of the reduced order and full order models without the confounding influence of noise. For each of the trials shown in Fig 6 the adaptive reduction accurately reflects the timing of action potentials for the full order model while the accuracy of the phase-only model tends to degrade over time.

4.3. Illustration in a model for the mammalian circadian clock

Here, the proposed model identification strategy is illustrated for a 16-dimensional computational model for the mammalian circadian clock comprised of interacting regulatory loops associated with Per, Cry, Bmal1 and Clock genes [42]. Full dynamical equations are given in Appendix B. The influence of light is included by modulating $v_{sp}$, which sets the maximum the rate of Per expression: $v_{sp}$ is chosen as the input in the implementation of the adaptive phase-isostable reduction (13). Here, $v_{sp} = 1.2$ nM/h is considered a state of total darkness and $v_{sp} = 1.8$ nM/h corresponds to a maximum level of brightness. Colored lines in Panel A of Fig. 7 shows periodic orbits that result for values $v_{sp} \in [1.2, 1.6]$. Note that while the full model has 16 state variables, only the mRNA concentrations of Per, Cry and Bmal1 ($M_p$, $M_c$, and $M_b$, respectively) are plotted here. Each periodic orbit has an unperturbed period $T(v_{sp}) \in [23.7, 24.2]$. In panel A the black line shows an example trajectory starting on the periodic orbit that results when taking $v_{sp} = 1.2$ nM/h. Over the course of the simulation, $v_{sp}$ is elevated to 1.8 nM/h for a duration of 60 h. The resulting trajectory quickly leaves the vicinity of the $v_{sp} = 1.2$ nM/h orbit, traveling closer to other orbits over the course of the simulation. Panels B and C highlight that the response of the phase and isostable coordinate to small step function perturbations is substantially different for each orbit indicating that the standard first order accurate phase reduction (5) will be unable to accurately capture the effects of these inputs.

For implementation of the adaptive phase-isostable reduced Eqs. (13), $\theta = 0$ for each orbit corresponds to the moment that the trajectory crosses Poincaré section for which $M_p = 1$ nM with a positive slope. Methods from [23] can be used to infer the terms $D(\theta, v_{sp})$ and $Q_2(\theta, v_{sp})$ from time series data assuming that traces of $M_p(t)$ can be accurately recorded. Note that this model

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**Fig. 6.** Panels A, B, and C compare transmembrane voltage outputs for the full neural model simulation from Eq. (48) (blue), adaptive reduction (red), and phase-only reduction (yellow) in response to the inputs $i_b(t)$ from panels G, H, and I, respectively. Panels D, E, and F show the error between these full and reduced order simulations over time. The horizontal dashed line at $\theta_{\text{red}} - \theta_{\text{act}} = 0$ is shown for reference. Green lines in panels G, H, and I show traces of the adaptive parameter $p(t)$ over the course of simulations of the adaptive reduction. For slower varying inputs, the adaptive parameter $p(t)$ follows $i_b(t)$ more closely.
Eq. (53), one finds $Q_\gamma$ where to the data. For each trial using the step function input from (56), an estimate of $\kappa_1(v_{sp}, g + \Delta p)$ can be obtained according to Eq. (19) by considering the relaxation to the periodic orbit $x_i(t_{i,0} + \Delta p)$ after the step function input that occurs at $t_p$.

4.3.1. Direct method inference of reduced order model terms using step function inputs

A model of the form (13) with a single isostable coordinate will be inferred from simulations of the circadian model from Appendix B in response to step function inputs. In contrast to the previous example, $D(\theta, v_{sp})$ and $Q_1(\theta, v_{sp})$ will be approximated using the direct method and subsequently used to infer the terms of $Z(\theta, v_{sp})$ and $I_1(\theta, v_{sp})$ according to Eqs. (38) and (40), respectively. To implement the direct method, consider an initial state $x_0 \in \mathcal{X}_{v_{sp}, 0}$ with a phase $\theta(x_0, v_{sp, 0}) = \theta_0$. Such an initial condition can be obtained by holding $v_{sp}$ at $v_{sp, 0}$ long enough for transients to decay. Letting $v_{sp}(t) = v_{sp, 0} + h(t_{p}) \Delta p$. 

where $h(t)$ is the Heaviside step function, the curve $D(\theta_0, v_{sp, 0})$ can be approximated according to $D(\theta_0, v_{sp, 0}) \approx \frac{2\pi n - \theta_0 - \alpha(t_{p, n} - t_p)}{\Delta p}$, 

where $t_{p, i}$ is the $i$th crossing of the Poincaré section that defines $\theta = 0$ after step function input is applied. In Eq. (57), $n$ must be chosen large enough so that $x_i(t_{p, n})$ is close to $x_i(v_{sp, 0} + \Delta p)$. Likewise, with the approximation $Q_1(\theta_0, v_{sp, 0})$.

\begin{equation}
Q_1(\theta_0, v_{sp, 0}) \approx \frac{\kappa_1(2\pi n - 1 - \alpha(t_{p, n} - t_{p, 1})))}{\Delta p(\exp(\kappa_1(t_{p, n} - t_p)) - \exp(\kappa_1(t_{p, 1} - t_p))))}.
\end{equation}

Note that in Eqs. (57) and (58), $\kappa_1$ and $\omega$ correspond to the Floquet exponent and natural frequency associated with the $x_i(v_{sp, 0} + \Delta p)$ orbit. These equations can be used to provide pointwise approximations for $D(\theta, p)$ and $Q_1(\theta, p)$ and a periodic function can be fit to the data. For each trial using the step function input from (56), an estimate of $\kappa_1(v_{sp, 0} + \Delta p)$ can be obtained according to Eq. (19) by considering the relaxation to the periodic orbit $x_i(v_{sp, 0} + \Delta p)$ after the step function input that occurs at $t_p$.

4.3.2. Reduced order model validation

Eqs. (57) and (58) are used to approximate $D(\theta, v_{sp})$ and $Q_1(\theta, v_{sp})$ results with shown as open circles in panels B and C of Fig. 7. Associated infinitesimal phase and isostable response curves are inferred using Eqs. (38) and (40) and shown as solid lines in Fig. 8 for different choices of $v_{sp}$. Open circles represent ground truth measurements and are obtained from simulations using the direct method with pulsatile perturbations instead of step function inputs. For simulations of the adaptive phase-isostable reduced order equations, the update rule $\dot{\psi} = \mathcal{G}(\psi, \theta, \psi_1) = -10^4 Q_1(\theta, \psi_2)$ of the same form as (16). For simulations of the standard phase reduced Eq. (5), infinitesimal phase response curves associated with the periodic orbit for which $v_{sp} = 1.2$ nM are used. For different initial conditions that correspond to different phases on the $v_{sp} = 1.2$ nM periodic orbit, Fig. 9 compares the predicted and actual phase shifts that result when taking $v_{sp}(t) = 1.2 + 0.3(h(0) - h(L))$ nM. Here where $L$ sets the duration of the pulse in $v_{sp}$. Panels A, B, and C show a single trial taking $L = 24$ hours with an initial condition of $\theta = 0$. Traces of $M_f(t)$ are shown in panel A. In response to the pulse, there is a slight mismatch in the timing of the outputs between the full model and the adaptive phase-isostable reduced order model; the phase-only reduction does not accurately reflect the response to the pulse. For the adaptive reduction, panel B shows the trace of $\psi(t)$ that results for the input shown in panel C. Panels D-G give representative results for initial conditions associated with different phases and for inputs of different duration. Here, $\Delta \theta \equiv \theta_f - \theta_0$, where $\theta_f$ is the phase resulting from the perturbation and $\theta_0$ is the phase that would have resulted had the perturbation not been applied. For short duration pulses, both the adaptive reduction and phase reduction strategies accurately predict the resulting phase shift. As the duration of the applied pulse is increased, the adaptive reduction strategy provides substantially more accurate results than the phase reduced equations.

5. Discussion and conclusion

Reduced order models based on the recently proposed adaptive phase-isostable reduction (13) are generally more accurate than models that use either standard phase reduction (2) or phase-isostable reduction (9) when large magnitude inputs are
Fig. 8. Solid lines in each panel show infinitesimal phase and isostable response curves from (13) inferred using only information about \( D(\theta, v_s) \) and \( Q_j(\theta, v_s) \) from panels B and C of Fig. 7. Open circles represent ground truth data points obtained using Eq. (3) with pulse inputs.

Fig. 9. For the circadian model, the input \( v_s \) is elevated from 1.2 nM (corresponding to total darkness) to 1.5 nM (corresponding to a moderate light intensity) for a finite duration of time. Panel A shows the response to a 24 h duration pulse simulated using the adaptive reduction (13), phase-only reduction (5), and the full model equations (B.1)–(B.16). For the simulation from panel A, Panel B shows the time course of the variable \( p \) from the adaptive reduction and the associated input is shown in Panel C. Panels D–G aggregate results for different initial conditions (with different phases, \( \theta \)) and for different pulse durations. On balance, both the adaptive reduction and phase reduction accurately predict the phase shift for short duration pulses in \( v_s \). For long duration pulses, the adaptive reduction approach substantially outperforms the standard phase reduction.

used. This work details a data-driven strategy to infer the necessary terms of the adaptive phase-isostable reduction from (13) which overcomes the limitations of these other reduction strategies highlighted in Fig. 1, i.e., that they do not accurately reflect system behavior when large magnitude inputs are applied.

The important implication from this work is that for a general system, provided it is possible to apply a series of either pulse inputs or step inputs and examine the relaxation to the underlying limit cycle, then it is also possible to obtain adaptive phase-isostable reduced order models of the form (13). As such, the proposed strategy requires no more information than the well-established direct method [2,9,10] and yields substantially more accurate reduced order models. By considering the change in the periodic orbit in response to static parameter perturbations in a basis of phase and isostable coordinates, the term of \( D(\theta, p) \) can ultimately be written as an explicit function of the infinitesimal phase response curve \( Z(\theta, p) \) using Eq. (34). Likewise, each \( Q_j(\theta, p) \) can be written as an explicit function of the associated infinitesimal isostable response curve \( I_j(\theta, p) \) using Eq. (35). In contrast to the approach suggested in [35], all the necessary terms of adaptive phase-isostable reduced equations can be obtained by either applying pulse inputs or step function inputs; using both types of inputs is not required and provides redundant information.
The proposed strategy is illustrated for three different models. In Section 4.1, a simple model with analytic solutions is considered to verify the derived relationships between the terms of the adaptive phase-isostable reduction. In Section 4.2, a series of pulse inputs is applied to infer the terms of the adaptive phase-isostable reduction in a conductance-based model neuron subject to noise in the transmembrane voltage variable. Section 4.3 considers only step function inputs to determine these terms for a computational model of the mammalian circadian clock. In each case, the adaptive reduction is substantially better at capturing the response to large magnitude inputs than the phase-only reduction.

The model identification strategy presented here relies on the adaptive phase-isostable reduction methodology. As such, it inherits the same limitations. For instance, it is only valid in regimes where the underlying model equations have oscillatory dynamics and cannot accommodate critical points of bifurcations in the allowable parameter sets. Additionally, the design of an appropriate parameter update rule $C_p$ is not always straightforward when multiple isostable coordinates are considered. Also, the terms from the reduction from Eq. (13) must vary continuously with respect to both $\theta$ and $p$. As such, the underlying system equations must have a sufficient degree of smoothness. Additional limitations of the adaptive phase-isostable reduction strategy are discussed in [31].

As a matter of practical implementation, there is still much room for improvement in strategies for inference of the infinitesimal phase and isostable response curves. For instance, the model inference strategies described in Sections 4.2.1 and 4.3.1 do not give perfect approximations of the required response curves and Floquet exponents. These approximation issues are compounded in noisy environments where there must be a balance between using an input that is large enough to discern a signal from the noisy data, but not so large as to drive the state too far away from the underlying limit cycle thereby invalidating the weak perturbation assumption. The development of more accurate strategies for data-driven inference of phase-isostable reduced order models of the form (9) would greatly benefit the adaptive phase-isostable model identification strategies presented here.

### Declaration of competing interest

The authors declare that they have no known competing financial interests or personal relationships that could have appeared to influence the work reported in this paper.

### Data availability

Data will be made available on request.

### Acknowledgment

This material is based upon work supported by the National Science Foundation Grant No. CMMI-2140527.

### Appendix A. Wang–Buzsáki model equations

Equations governing the Wang–Buzsáki model [40] considered in Section 4.2 are given below:

\[ CV = -g_n h n^3 (V - E_Na) - g_l n^4 (V - E_L) - g_l (V - E_L) - i_i + i_n, \]
\[ h = \gamma \{ a_1 [1 - n - \beta_1 (V)] n \}, \]
\[ n = \gamma \{ a_2 [1 - n - \beta_2 (V)] n \}, \]
\[ i_i = \alpha (1.5 / (1 + \exp(b - V/k)) - w). \]  
(A.1)

Above, $V$ represents the transmembrane voltage and $h$ and $n$ are gating variables. This model is augmented with an adaptation current $[41], i_w$, which is governed by the variable $w$. The baseline current, $i_n$, is chosen as the adaptive parameter $p$. The membrane capacitance, $C = 1 \mu F/cm^2$. Auxiliary equations are

\[ m_\infty(V) = \alpha_m(V) / (\alpha_m(V) + \beta_m(V)); \]
\[ \beta_m(V) = 0.125 \exp(- (V + 44)/80); \]
\[ \alpha_a (V) = -0.01 (V + 34) / (\exp(-0.1(V + 34)) - 1); \]
\[ \beta_a (V) = 1 / (\exp(-0.1(V + 28)) + 1); \]
\[ \alpha_h (V) = 0.07 \exp(- (V + 58)/20); \]
\[ \beta_h (V) = 4 \exp(- (V + 60)/18); \]
\[ \alpha_r (V) = -0.1 (V + 35) / (\exp(-0.1(V + 35)) - 1). \]

Reversal potentials and conductances are taken to be $E_{Na} = 55 \text{ mV}, E_K = -90 \text{ mV}, E_L = -65 \text{ mV}, g_{Na} = 35 \text{ mS/cm}^2, g_K = 9 \text{ mS/cm}^2, g_l = 0.1 \text{ mS/cm}^2, \gamma = 5$ influences the rate of change of the gating variables. The adaptation current is given by

\[ i_w = g_w w (V - E_K). \]

Parameters associated with the adaptation current are $a = 0.02 \text{ m s}^{-1}, b = -5 \text{ mV}, k = 0.5 \text{ mV}, g_w = 2 \text{ ms/cm}^2$.

### Appendix B. Circadian model equations

The equations for the circadian model considered in Section 4.3 were presented in [42]. The model contains 16 variables. The mRNA concentrations of Per, Cry, and Bmal1 are denoted by $M_P, M_C$, and $M_B$, respectively. Phosphorylated (resp. nonphosphorylated) Per and Cry proteins in cytosol are denoted by $P_{CP}$ and $C_{CP}$ (resp., $P_{C}$ and $C_{C}$). Concentrations of Per-Cry complex in cytosol and nucleus are denoted by $P_{C}C_{N}, P_{CP}C_{N}, P_{CP}C_{P}$, and $P_{CP}C_{CP}$; Concentrations of Bmal1 in cytosol and nucleus are denoted by $B_{C}, B_{CP}, B_{CN},$ and $B_{CP}$. Here, the subscripts $C, N, CP$ and $NP$ denote cytosolic, nuclear, cytosolic phosphorylated, and phosphorylated forms, respectively. Finally, $I_N$ denotes the inactive complex between Per-Cry and Clock-Bmal1 in the nucleus.

The model equations that govern the dynamics are:

\[ \dot{M}_P = v_P (f) \frac{B_{mP}^n}{K_{mP} + B_{mP}^n} - v_{mP} \frac{M_P}{K_{vP} + M_P} - k_{dMP} M_P, \]  
(B.1)

\[ \dot{M}_C = v_C \frac{K_{mC} + B_{mC}^n}{B_{mC} + B_{mC}^n} - v_{mC} \frac{M_C}{K_{vC} + M_C} - k_{dMC} M_C, \]  
(B.2)

\[ \dot{M}_B = v_B \frac{K_{mB} + B_{mB}^n}{B_{mB} + B_{mB}^n} - v_{mB} \frac{M_B}{K_{vB} + M_B} - k_{dMB} M_B, \]  
(B.3)

\[ \dot{P}_C = k_{dPC} M_P - V_{1P} \frac{P_C}{K_p + P_C} + V_{2P} \frac{P_{CP}}{K_d + P_{CP}}, \]  
(B.4)

\[ \dot{C}_C = k_{dCC} M_C - V_{1C} \frac{C_C}{K_p + C_C} + V_{2C} \frac{C_{CP}}{K_d + C_{CP}}, \]  
(B.5)

\[ \dot{P}_{CP} = V_{1P} \frac{P_C}{K_p + P_C} + V_{2P} \frac{P_{CP}}{K_d + P_{CP}} - v_{dPC} \frac{P_{CP}}{K_d + P_{CP}} - k_{dPC} P_{CP}. \]  
(B.6)

\[ \dot{C}_{CP} = V_{1C} \frac{C_C}{K_p + C_C} - V_{2C} \frac{C_{CP}}{K_d + C_{CP}} - v_{dCC} \frac{C_{CP}}{K_d + C_{CP}} - k_{dCP} C_{CP}, \]  
(B.7)

\[ \dot{P}_{CC} = -V_{1PC} \frac{P_{CC}}{K_p + P_{CC}} + V_{2PC} \frac{P_{CP}C_{P}}{K_d + P_{CP}} - k_{dPC} P_{CC} + k_{3P} C_{C}, \]  
(B.8)
Appendix D. Identification of reduced order model parameters using operational phase coordinates

As discussed in Section 2.4, operational phase coordinates can be used to obtain information about transient behavior associated with the decay of solutions to an underlying stable limit cycle. In this work, results from [23] are leveraged to accomplish this task, with useful results summarized in Eqs. (17)–(19). This appendix is provided to clearly illustrate how Eqs. (17)–(19) from the main text can be obtained using results that were previously presented in [23].

In this work, \( \theta \) and \( \theta^* \) represent asymptotic and operational phase coordinates, respectively, for a stable limit cycle \( x_p^* \) that emerges in Eq. (1) when \( u = p \) is held constant. As illustrated in Eq. (15) from Ref. [23], if \( u \) is held at \( p \), when considering a single isostable coordinate \( \psi \) in the reduced order model, to leading order \( \epsilon \) the time between the \( k \)th crossing of the \( \theta = 0 \) and \( \theta^* = 0 \) level set is related by

\[
    t_k^* = t_k - \frac{\log(\alpha_1)}{n \cdot x_p^*|_{\theta=0}}.
\]

where \( t_k^* \) corresponds to the \( k \)th crossing of the \( \theta^* = 0 \) level set, \( t_k \) corresponds to the \( k \)th crossing of the \( \theta = 0 \) level set, \( g_k(0, p) \) is the Floquet eigenfunction associated with \( \psi \) isostable coordinate evaluated at \( \theta = 0, \) \( x_p^*|_{\theta=0} \) is the time derivative of the periodic orbit evaluated at \( \theta = 0, \) \( n \) is a unit vector normal to the \( \theta^* = 0 \) level set, and the dot denotes the dot product. Additionally, Ref. [23] defines the following constant

\[
    \alpha_1 = \frac{\gamma g_k(0, p)(n \cdot g_k(0, p))}{n \cdot x_p^*|_{\theta=0}}.
\]

Note that \( g_k(\theta, p) \) is a Floquet eigenfunction which is unique to a constant scaling; in this work, it is assumed that \( n \cdot x_p^*|_{\theta=0} \neq 0 \) so that the normalization \( \alpha_1 = 1 \) can be mandated with an appropriate scaling of \( g_k(\theta, p) \). Combining Eqs. (D.1) and (D.2) yields Eq. (17) from the main text. Eq. (18) from the main text can be obtained through algebraic manipulation of Eq. (D.2) recalling the scaling \( \alpha_1 = 1 \) and noting that \( \frac{\partial}{\partial \tau} = \omega \frac{\partial}{\partial \theta} \). Eq. (19) from the main text can be derived starting with Equation (39) from [23]:

\[
    \log(\tau_n) = \alpha + n \log(\lambda_1),
\]

where \( \tau_n = t_{n+1}^* - t_n^* - T(p) \), where \( t_{n+1}^* \) is the \( n \)th crossing of the \( \theta^* = 0 \) level set when \( u \) is held constant at \( p, \) \( \alpha \) is a constant that depends on system parameters and initial conditions, \( T(p) = 2\pi/\omega(p) \) is the period of the limit cycle, and \( \lambda_1 \) is the Floquet multiplier associated with the isostable coordinate \( \psi. \) Considering Eq. (D.3) at successive indices \( n \) and \( n + 1, \) through direct manipulation one finds

\[
    \log(\lambda_1) = \log(\tau_{n+1}/\tau_n).
\]

Eq. (19) follows directly from the relation \( \kappa_1(p) = \log(\lambda_1)/T(p) \) which is a standard relationship between Floquet exponents and Floquet multipliers.

References